MODEL PREDICTIVE CONTROL FOR CYBER-PHYSICAL SYSTEMS

Alberto Bemporad

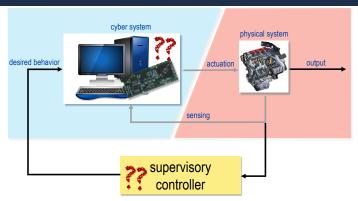
imt.lu/ab





(extended slide set: http://cse.lab.imtlucca.it/~bemporad/mpc_course.html)

CONTROL ISSUES IN CYBER-PHYSICAL SYSTEMS



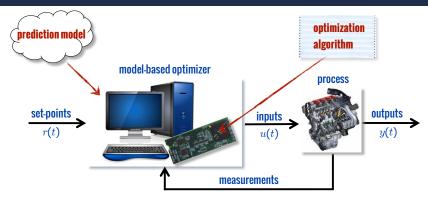
- What to embed inside the cyber system to make the physical system behave autonomously in a robust, safe, and optimal manner?
- How to synthesize a supervisory controller for a CPS?

CONTENTS OF MY LECTURE

- Model predictive control for CPS's
- Embedded quadratic optimization algorithms (inside the CPS)
- Hybrid MPC = supervisory control of CPS's
- Data-driven controller synthesis for CPS's



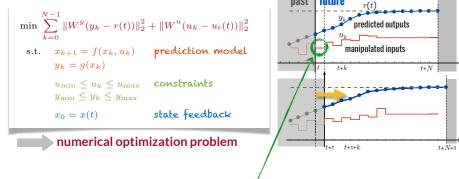
MODEL PREDICTIVE CONTROL (MPC)



Use a dynamical model of the process to predict its future evolution and choose the "best" control action

MODEL PREDICTIVE CONTROL (MPC)

ullet Goal: find the best control sequence over a future horizon of N steps



• At each time t:

- get new measurements to update the estimate of the current state x(t)
- solve the optimization problem with respect to $\{u_0,\ldots,u_{N-1}\}$
- apply only the first optimal move $u(t)=u_0^st,$ discard the remaining samples

MPC IN INDUSTRY

The MPC concept for process control dates back to the 60's

Discrete Dynamic Optimization
Applied to On-Line Optimal Control



MARSHALL D. RAFAL and WILLIAM F. STEVENS

(Rafal, Stevens, AiChE Journal, 1968)

• MPC used in the process industries since the 80's

(Qin, Badgewell, 2003) (Bauer, Craig, 2008)



Today APC (advanced process control) = MPC

• Impact of advanced control technologies in industry

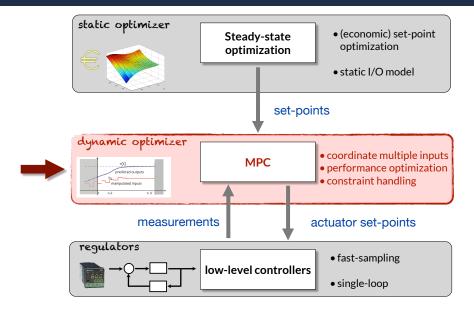
TABLE 1 A list of the survey results in order of industry impact as perceived by the committee members.			
Rank and Technology	High-Impact Ratings	Low- or No-Impact Ratings	
PID control	100%	0%	
Model predictive control	78%	9%	
System identification	61%	9%	
Process data analytics	61%	17%	
Soft sensing	52%	22%	
Fault detection and identification	50%	18%	
Decentralized and/or coordinated control	48%	30%	
Intelligent control	35%	30%	
Discrete-event systems	23%	32%	
Nonlinear control	22%	35%	
Adaptive control	17%	43%	
Robust control	13%	43%	
Hybrid dynamical systems	13%	43%	

MPC IN INDUSTRY

(Samad, IFAC Newsletter, April 2019)

	Current Impact	Future Impact
Control Technology	% High Low/No	High Low/No
PID control	91% 0%	78% 6%
System Identification	65% 5%	72% 5%
Estimation & filtering	64% 11%	63% 3%
Model-predictive contro	62% 11%	85% 2%
Process data analytics	51% 15%	70% 8%
Fault detection &	48% 17%	8% 8%
identification		
Decentralized and/or	29% 33%	54% 11%
coordinated control		
Robust control	26% 35%	42% 23%
Intelligent control	24% 38%	5 9% 11%
Nonlinear control	21% 44%	42% 15%
Discrete-event systems	24% 45%	39% 27%
Adaptive control	18% 38%	44% 17%
Repetitive control	12% 74%	17% 51%
Other advanced	11% 64%	25% 39%
control technology		
Hybrid dynamical	11% 68%	33% 33%
systems		
Game theory	5% 76%	17% 52%

TYPICAL USE OF MPC



MPC OF AUTOMOTIVE SYSTEMS

(Bemporad, Bernardini, Borrelli, Cimini, Di Cairano, Esen, Giorgetti, Graf-Plessen, Hrovat, Kolmanovsky Levijoki, Livshiz, Long, Pattipati, Ripaccioli, Trimboli, Tseng, Verdejo, Yanakiev, ..., 2001-present)

Powertrain

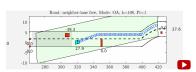
engine control, magnetic actuators, robotized gearbox, power MGT in HEVs, cabin heat control, electrical motors

Vehicle dynamics

traction control, active steering, semiactive suspensions, autonomous driving

Ford Motor Company
Jaguar
DENSO Automotive
FCA
General Motors



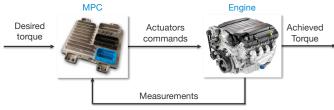


Most automotive OEMs are looking into MPC solutions today

ODYS

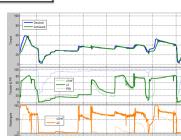
MPC OF GASOLINE TURBOCHARGED ENGINES

Control throttle, wastegate, intake & exhaust cams to make engine torque track set-points, with max efficiency and satisfying constraints



numerical optimization problem solved in real-time on ECU

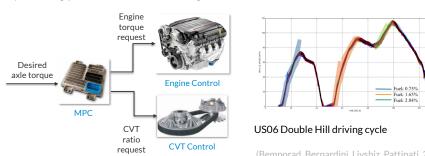
(Bemporad, Bernardini, Long, Verdejo, 2018)



engine operating at low pressure (66 kPa)

SUPERVISORY MPC OF POWERTRAIN WITH CVT

- Coordinate engine torque request and continuously variable transmission (CVT) ratio to improve fuel economy and drivability
- Real-time MPC is able to take into account coupled dynamics and constraints, optimizing performance also during transients



MPC IN AERONAUTIC INDUSTRY

PRESS RELEASE

Pratt & Whitney's F135 Advanced Multi-Variable Control Team Receives UTC's Prestigious George Mead Award for Outstanding Engineering Accomplishment



Pratt & Whitney engineers Louis Celiberti, Timothy Crowley, James Fuller and Cary Powell won the George Mead Award – United Technologies Corp.'s highest award for outstanding engineering achievement – for their pioneering work in developing the world's first advanced multi-variable control (AWVC) design for the only engine that powers the F-35 Lightning II flight test program. Pratt & Whitney is a United Technologies Corp. (NYSE:UTX) company.

The AMVC, which uses a proprietary model predictive control methodology, is the most technically advanced propulsion system control ever produced by the aerospace industry, demonstrating the highest pilot rating for flight performance and providing independent control of vertical thrust and pitch from five sources. This innovative and industry-leading advanced design is protected with five broad patents for Pratt & Whitney and UTC, and is the new standard for propulsion system control for Pratt & Whitney military and commercial engines.





http://www.pw.utc.com/Press/Story/20100527-0100/2010

OTHER EXAMPLES OF MPC APPLICATIONS

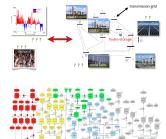
MPC for smart electricity grids

Example: Dispatch power in smart distribution grids, trade energy on energy markets

MPC of drinking water networks

Example: save \approx 5% energy costs in Barcelona's drinking water network w.r.t. current practice

MPC for financial engineering Example: dynamic portfolio optimization for option hedging







All the above applications require stochastic MPC formulations



LINEAR MPC

• Linear prediction model:
$$\begin{cases} x_{k+1} &= Ax_k + Bu_k \\ y_k &= Cx_k \end{cases}$$

$$\begin{aligned} x &\in \mathbb{R}^n \\ u &\in \mathbb{R}^m \\ y &\in \mathbb{R}^p \end{aligned}$$

• Constraints to enforce:

$$\begin{cases} u_{\min} \le u(t) \le u_{\max} \\ y_{\min} \le y(t) \le y_{\max} \end{cases}$$

Constrained optimal control problem (quadratic performance index):

$$\min_{z} x'_{N} P x_{N} + \sum_{k=0}^{N-1} x'_{k} Q x_{k} + u'_{k} R u_{k}$$

s.t.
$$u_{\min} \le u_k \le u_{\max}, k = 0, ..., N - 1$$

 $y_{\min} \le y_k \le y_{\max}, k = 1, ..., N$

$$\begin{vmatrix} R & = & R' \succ 0 \\ Q & = & Q' \succeq 0 \\ P & = & P' \succeq 0 \end{vmatrix} z = \begin{bmatrix} u_0 \\ u_1 \\ \vdots \\ u_{N-1} \end{bmatrix}$$

LINEAR MPC

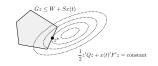
- Linear prediction model: $x_k = A^k x_0 + \sum_{i=0}^{\kappa-1} A^i B u_{k-1-i}$
- Optimization problem (condensed form):

$$V(x_0)=rac{1}{2}x_0'Yx_0+ \min_z rac{1}{2}z'Hz+x_0'F'z$$
 (quadratic objective)

$${
m s.t.} \quad Gz\leq W+Sx_0 \quad \hbox{(linear constraints)}$$

convex Quadratic Program (QP)

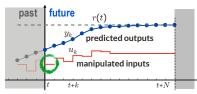
•
$$z=\left|\begin{array}{c} u_0\\u_1\\ \vdots\\u_{N-1}\end{array}\right|\in\mathbb{R}^{Nm}$$
 is the optimization vector



• $H=H'\succ 0$, and H,F,Y,G,W,S depend on weights Q,R,P upper and lower bounds $u_{\min},u_{\max},y_{\min},y_{\max}$ and model matrices A,B,C.

LINEAR MPC ALGORITHM

@ each sampling step t:



Measure (or estimate) the current state x(t)

$$\bullet \ \ \text{Get the solution} \ z^* = \left[\begin{array}{c} u_0^* \\ u_1^* \\ \vdots \\ u_{N-1}^* \end{array} \right] \ \text{of the QP}$$

$$\bullet \ \ \text{Get the solution} \ z^* = \begin{bmatrix} u_0^* \\ u_1^* \\ \vdots \\ u_{N-1}^* \end{bmatrix} \ \text{of the QP} \qquad \begin{cases} & \underbrace{\min}_{z} \quad \frac{1}{2}z'Hz + \overbrace{x'(t)F'z} \\ & \text{s.t.} \quad Gz \leq W + S \underbrace{x(t)}_{\text{feedback}} \end{cases}$$

• Apply only $u(t)=u_0^*$, discarding the remaining optimal inputs u_1^*,\dots,u_{N-1}^*

(Keerthi, Gilbert, 1988) (Bemporad, Chisci, Mosca, 1994)

• Theorem: Let the MPC law be based on

$$V^*(x(t)) = \min \qquad \sum_{k=0}^{N-1} x_k' Q x_k + u_k' R u_k$$
 s.t.
$$x_{k+1} = A x_k + B u_k$$

$$u_{\min} \le u_k \le u_{\max}$$

$$y_{\min} \le C x_k \le y_{\max}$$

$$x_N = 0 \qquad \leftarrow \text{"terminal constraint"}$$

with $R,Q \succ 0$, $u_{\min} < 0 < u_{\max}$, $y_{\min} < 0 < y_{\max}$. If the optimization problem is feasible at time t=0 then

$$\lim_{t \to \infty} x(t) = 0, \quad \lim_{t \to \infty} u(t) = 0$$

and the constraints are satisfied at all time $t \geq 0$, for all $R, Q \succ 0$.

• Many more convergence and stability results exist (Mayne, 2014)

LINEAR MPC - TRACKING

Optimal control problem (quadratic performance index):

$$\min_{z} \sum_{k=0}^{N-1} \|W^{y}(y_{k+1} - r(t))\|_{2}^{2} + \|W^{\Delta u} \Delta u_{k}\|_{2}^{2}$$

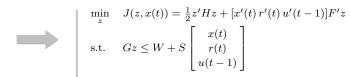
$$[\Delta u_{k} \triangleq u_{k} - u_{k-1}], u_{-1} = u(t-1)$$
s.t. $u_{\min} \leq u_{k} \leq u_{\max}, k = 0, \dots, N-1$

$$y_{\min} \leq y_{k} \leq y_{\max}, k = 1, \dots, N$$

$$\Delta u_{\min} \leq \Delta u_{k} \leq \Delta u_{\max}, k = 0, \dots, N-1$$

$$z = \begin{bmatrix} \Delta u_{0} \\ \Delta u_{1} \\ \vdots \\ \Delta u_{N-1} \end{bmatrix}$$
or $z = \begin{bmatrix} u_{0} \\ u_{1} \\ \vdots \\ u_{N-1} \end{bmatrix}$

weight W = diagonal matrix (more generally, Cholesky factor of Q=W'W)



convex Quadratic Program

- Add the extra penalty $\|W^u(u_k-u_{\mathrm{ref}}(t))\|_2^2$ to track input references
- Constraints may depend on r(t), such as $e_{\min} \leq y_k r(t) \leq e_{\max}$

INTEGRAL ACTION AND $\triangle u$ -formulation

 In control systems, integral action occurs if the controller has a transfer-function from the output to the input of the form

$$u(t) = \frac{B(z)}{(z-1)A(z)}y(t), \qquad B(1) \neq 0$$

• One may think that the Δu -formulation of MPC provides integral action ...

... is it true?

 \bullet $\,$ Example: we want to regulate the output y(t) to zero of the scalar system

$$x(t+1) = \alpha x(t) + \beta u(t)$$

$$y(t) = x(t)$$

INTEGRAL ACTION AND Δu -formulation

 $\bullet \;\;$ Design an unconstrained MPC controller with horizon N=1

$$\Delta u(t) = \arg \min_{\Delta u_0} \Delta u_0^2 + \rho y_1^2$$
s.t. $\Delta u_0 = u_0 - u(t-1)$

$$y_1 = x_1 = \alpha x(t) + \beta(\Delta u_0 + u(t-1))$$

• By substitution, we get

$$\Delta u(t) = \arg \min_{\Delta u_0} \Delta u_0^2 + \rho(\alpha x(t) + \beta u(t-1) + \beta \Delta u_0)^2
= \arg \min_{\Delta u_0} (1 + \rho \beta^2) \Delta u_0^2 + 2\beta \rho(\alpha x(t) + \beta u(t-1)) \Delta u_0
= -\frac{\beta \rho \alpha}{1 + \rho \beta^2} x(t) - \frac{\rho \beta^2}{1 + \rho \beta^2} u(t-1)$$

• Since x(t) = y(t) and $u(t) = u(t-1) + \Delta u(t)$ we get the linear controller

$$u(t) = \frac{\frac{\rho\beta\alpha}{1+\rho\beta^2}z}{z - \frac{1}{1+\rho\beta^2}}y(t)$$
 no pole in z=1

• Reason: MPC gives a feedback gain on both x(t) and u(t-1), not just on x(t)

OUTPUT INTEGRATORS AND OFFSET-FREE TRACKING

Add constant unknown disturbances on measured outputs:

$$\begin{cases} x_{k+1} = Ax_k + Bu_k \\ d_{k+1} = d_k \\ y_k = Cx_k + d_k \end{cases}$$

- Use the extended model to design a state observer (e.g., Kalman filter) that estimates both the state $\hat{x}(t)$ and disturbance $\hat{d}(t)$ from y(t)
- Why we get offset-free tracking in steady-state (intuitively):
 - the observer makes $C\hat{x}(t) + \hat{d}(t) \rightarrow y(t)$
 - the MPC controller makes $C\hat{x}(t)+\hat{d}(t) \rightarrow r(t)$ (predicted tracking error)
 - the combination of the two makes $y(t) \rightarrow r(t)$

(estimation error)

(actual tracking error)

- In steady state, the term $\hat{d}(t)$ compensates for model mismatch
- See more on survey paper (Pannocchia, Gabiccini, Artoni, 2015)

ANTICIPATIVE ACTION (A.K.A. "PREVIEW")

$$\min_{\Delta U} \sum_{k=0}^{N-1} \|W^{y}(y_{k+1} - r_{k+1})\|_{2}^{2} + \|W^{\Delta u}\Delta u(k)\|_{2}^{2}$$

 Reference not known in advance (causal):

$$r_k \equiv r(t), \forall k = 0, \dots, N-1$$
 Output I reference
$$r(t)$$
 0.5 10 15 Input
$$r(t)$$
 0.5 10 15 Input
$$r(t)$$
 1 15 Input
$$r(t)$$

 Future refs (partially) known in advance (anticipative action):

$$r_k = r(t+k), \forall k = 0, \dots, N-1$$
 Output / reference use $r(t+k)$ of some state of the state of t

go to demo mpcpreview.m (MPC Toolbox)

• Same idea also applies for preview of measured disturbances

LTV-MPC

Linear Time-Varying (LTV) model predictive control

$$\begin{cases} x_{k+1} = A_k(t)x_k + B_k(t)u_k \\ y_k = C_k(t)x_k \end{cases}$$

- ullet The model can change at each time t, even over the prediction horizon k
- The resulting optimization problem is still a QP

$$\min_{z} \frac{1}{2}z'H(t)z + \begin{bmatrix} \frac{x(t)}{r(t)} \\ \frac{x(t)}{u(t-1)} \end{bmatrix}' F(t)'z$$
s.t.
$$G(t)z \le W(t) + S(t) \begin{bmatrix} \frac{x(t)}{r(t)} \\ \frac{x(t)}{u(t-1)} \end{bmatrix}$$

In LTV-MPC the QP matrices must be constructed online

LINEARIZING A NONLINEAR MODEL

LTV models can be obtained by linearizing nonlinear models

$$\begin{cases} \frac{dx_c(t)}{dt} &= f(x_c(t), u_c(t), p_c(t)) \\ y_c(t) &= g(x_c(t), p_c(t)) \end{cases}$$

• At time t, consider nominal trajectories

$$\begin{array}{lcl} U & = & \big\{\bar{u}_c(t), \bar{u}_c(t+T_s), \ldots, \bar{u}_c(t+(N-1)T_s)\big\} \\ & & \text{(example: } U \text{ = shifted previous optimal sequence or input ref. trajectory)} \\ P & = & \big\{\bar{p}_c(t), \bar{p}_c(t+T_s), \ldots, \bar{p}_c(t+(N-1)T_s)\big\} \end{array}$$

• Integrate the model and get nominal state/output trajectories

(no preview: $\bar{p}_c(t+k) \equiv \bar{p}_c(t)$)

$$X = \{\bar{x}_c(t), \bar{x}_c(t+T_s), \dots, \bar{x}_c(t+(N-1)T_s)\}$$

$$Y = \{\bar{y}_c(t), \bar{y}_c(t+T_s), \dots, \bar{y}_c(t+(N-1)T_s)\}$$

• Examples: $\bar{x}_c(t) = \text{current state / equilibrium state / reference state}$

LINEARIZATION AND TIME-DISCRETIZATION

• Getting the discrete-time LTV model $A_k(t)$, $B_k(t)$, $C_k(t)$ requires to linearize and discretize in time the nonlinear continuous-time dynamical model

$$\frac{dx_c(t)}{dt} = f(x_c, u_c, p_c) \approx \underbrace{f(\bar{x}_c, \bar{u}_c, \bar{p}_c)}_{\underbrace{d\bar{x}_c}_{dt}} + \underbrace{\frac{\partial f}{\partial x_c}\bigg|_{\bar{x}_c, \bar{u}_c, \bar{p}_c}}_{\underbrace{J_{ac}b_{ian} \text{ matrix } A_c}} \underbrace{\frac{\partial f}{\partial u_c}\bigg|_{\bar{x}_c, \bar{u}_c, \bar{p}_c}}_{\underbrace{J_{ac}b_{ian} \text{ matrix } B_c}}$$

• Let $x=x_c-\bar{x}_c$, $u=u_c-\bar{u}_c$. We get the continuous-time linear system

$$\frac{dx}{dt} = A_c x + B_c u$$

- Similarly, from the output equation we get $y=y_c-\bar{y}_cpprox \underbrace{\left.\frac{\partial g}{\partial x_c}\right|_{\bar{x}_c,\bar{u}_c,\bar{p}_c}}_{\text{Jacobian matrix }C}$
- Convert (A_c, B_c, C) to discrete-time model (A, B, C) (Euler method, exp. matrix, ...)
- LTV-MPC: @each time t simulate the NL model, get linearized models, build & solve QP

FROM LTV-MPC TO NONLINEAR MPC

(Gros, Zanon, Quirynen, Bemporad, Diehl, 2016)

NL-MPC: We can solve a sequence of LTV-MPC problems at each time t

For h = 0 to $h_{\text{max}} - 1$ do:

- 1. Simulate from x(t) with inputs U_h and get state trajectory X_h
- 2. Linearize around (X_h, U_h) and discretize in time
- 3. Get U_{h+1}^* = **QP solution** of corresponding LTV-MPC problem
- 4. Line search: find optimal step size $\alpha_h \in (0, 1]$;
- 5. Set $U_{h+1} = (1 \alpha_h)U_h + \alpha_h U_{h+1}^*$;

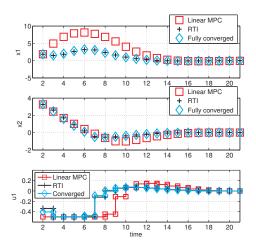
Return solution $U_{h_{\max}}$

- The above method is Sequential Quadratic Programming (SQP) applied to solve the full nonlinear MPC problem
- Special case: just solve one iteration with $\alpha=1$ (a.k.a. Real-Time Iteration) (Diehl, Bock, Schloder, Findeisen, Nagy, Allgower, 2002) = LTV-MPC

NONLINEAR MPC

(Gros, Zanon, Quirynen, Bemporad, Diehl, 2016)

• Example



PREDICTION MODELS FOR MPC

- Physics-based nonlinear models
- Use black-box system identification algorithms to fit linear or nonlinear models to data
- Use machine-learning techniques to get nonlinear models (such neural networks) from data, with Jacobians
- A mix of the above (gray-box models)
- Note: Computation complexity depends on chosen model, need to trade off descriptiveness vs simplicity of the model





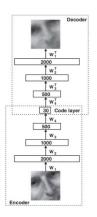


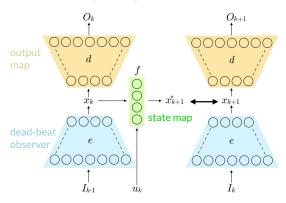


LEARNING NONLINEAR MODELS FOR MPC

Masti, Bemporad, 2018)

Idea: use autoencoders and artificial neural networks to learn a nonlinear state-space model of desired order from input/output data





ANN with hourglass structure

(Hinton, Salakhutdinov, 2006

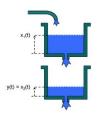
$$O_k = [y'_k \dots y'_{k-m}]'$$

 $I_k = [y'_k \dots y'_{k-n_a+1} u'_k \dots u'_{k-n_b+1}]'$

LEARNING NONLINEAR MODELS FOR MPC - AN EXAMPLE

lasti, Bemporad, 2018)

• System generating the data = nonlinear 2-tank benchmark

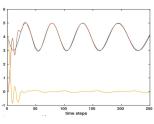


www.mathworks.com

$$\begin{cases} x_1(k+1) = x_1(k) - k_1\sqrt{x_1(k)} + k_2(u(k) + w(k)) \\ x_2(k+1) = x_2(k) + k_3\sqrt{x_1(k)} - k_4\sqrt{x_2(k)} \\ y(k) = x_2(k) + v(k) \end{cases}$$

Model is totally unknown to learning algorithm

- Artificial neural network (ANN): 3 hidden layers 60 exponential linear unit (ELU) neurons
- For given number of model parameters, autoencoder approach is superior to NNARX
- Jacobians directly obtained from ANN structure for Kalman filtering & MPC problem construction



LTV-MPC results

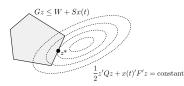


EMBEDDED MPC AND QUADRATIC PROGRAMMING

MPC based on linear models requires solving a Quadratic Program (QP)

$$\min_{z} \frac{1}{2}z'Qz + x'(t)F'z$$
 s.t.
$$Gz \le W + Sx(t)$$

$$z = \begin{bmatrix} u_0 \\ u_1 \\ \vdots \\ u_{N-1} \end{bmatrix}$$



On Minimizing A Convex Function Subject to Linear Inequalities $\mathbf{By} \,\, \mathbf{E}. \,\, \mathbf{M}. \,\, \mathbf{L}. \,\, \mathbf{Beale}$

Admiralty Research Laboratory, Teddington, Middlesex

The minimization of a convex function of variables subject to linear inequalities is discussed briefly in general terms. Dantizje Simplew Method is extended to yield finite algorithms for minimizing either a convex quadratic function or the sum of the / largest of a set of linear functions, and the solution of a generalization of the latter problem is indicated. In the last two sections a form of linear programmings from of a convex function.

(Beale, 1955)

A rich set of good QP algorithms is available today



• Not all QP algorithms are suitable for industrial embedded control

MPC IN A PRODUCTION ENVIRONMENT

Key requirements for deploying MPC in production:



- speed (throughput)
 - worst-case execution time less than sampling interval
 - also fast on average (to free the processor to execute other tasks)



- 2. limited memory and CPU power (e.g., 150 MHz / 50 kB)
- 3. numerical robustness (single precision arithmetic)





- 4. **certification** of worst-case execution time
- code simple enough to be validated/verified/certified (library-free C code, easy to check by production engineers)





EMBEDDED SOLVERS IN INDUSTRIAL PRODUCTION

- Multivariable MPC controller
- Sampling frequency = 40 Hz (= 1 QP solved every 25 ms)
- Vehicle operating \approx 1 hr/day for \approx 360 days/year on average
- Controller running on 10 million vehicles

~520,000,000,000,000 QP/yr and none of them should fail.



DUAL GRADIENT PROJECTION FOR QP

• Consider the strictly convex QP and its dual

$$\begin{array}{lll}
\min & \frac{1}{2}z'Qz + x'F'z \\
\text{s.t.} & Gz \le W + Sx
\end{array}
\qquad \begin{array}{lll}
\min & \frac{1}{2}y'Hy + (Dx + W)'y \\
\text{s.t.} & y \ge 0
\end{array}$$

with
$$H = GQ^{-1}G'$$
 , $D = S + GQ^{-1}F$. Take $L \geq \frac{1}{\lambda_{\max}(H)}$

Apply proximal gradient method to dual QP: (Combettes, Waijs, 2005)

$$y^{k+1} = \max\{y^k - \frac{1}{L}(Hy^k + Dx + W), 0\}$$
 $y_0 = 0$

- Primal solution: $z^k = -Q^{-1}(Fx + G'y^k)$
- Also works in fixed-point arithmetic (Patrinos, Guiggiani, Bemporad, 2015)
- Convergence is slow: the initial error $f(z^0) f(z^st)$ reduces as 1/k

• Solve (dual) QP by fast gradient method

$$\min_{z} \frac{1}{2}z'Qz + x'F'z$$
s.t.
$$Gz \le W + Sx$$

$$K = Q^{-1}G'$$

$$g = Q^{-1}Fx$$

$$L \ge \frac{1}{\lambda_{\max}(GQ^{-1}G')}$$

$$\beta_k = \max\{\frac{k-1}{k+2}, 0\}$$

$$w^{k} = y^{k} + \beta_{k}(y^{k} - y^{k-1})$$

$$z^{k} = -Kw^{k} - g$$

$$s^{k} = \frac{1}{L}Gz^{k} - \frac{1}{L}(W + Sx)$$

$$y^{k+1} = \max\{w^{k} + s^{k}, 0\}$$

```
while kcmaxiter
beta-max((k-1)/(k+2),0);
u=/yeta*(y-y0);
z=(iMC*w=iMC);
s=(0.72-b.;
y0=y;

* Termination
if all(s<-eps6L)
gapl==""s;
if gapl<-epsVL
return
end
end
y=m+5;
k=k+1;</pre>
```

- Very simple to code
- Convergence rate: $f(x^k) f(x^*) \le \frac{2L}{(k+2)^2} ||z_0 z^*||_2^2$ (Necoara, Nesterov, Glineur, 2018)
- Tight bounds on maximum number of iterations can be computed

(Gabay, Mercier, 1976) (Glowinski, Marrocco, 1975) (Douglas, Rachford, 1956) (Boyd et al., 2010)

 $\bullet \ \ \text{Alternating Directions Method of Multipliers for QP}$

$$\begin{array}{ll} \min & \frac{1}{2}z'Qz + c'z \\ \text{s.t.} & \ell \leq Az \leq u \end{array}$$

$$\begin{array}{lcl} z^{k+1} & = & -(Q+\rho A'A)^{-1}(\rho A'(u^k-s^k)+c) \\ s^{k+1} & = & \min\{\max\{Az^{k+1}+u^k,\ell\},u\} \\ u^{k+1} & = & u^k+Ax^{k+1}-s^{k+1} \end{array}$$

while k<maxiter
k=k+1;
z=-iM*(c+A'*(rho*(u-s)));
Az=A'z;
s=max(min(Az+u,ub),lb);
u=u+Az-S;
end</pre>

(7 lines EML code) (\approx 40 lines of C code)

 ρu = dual vector

- Very simple to code
- Sensitive to matrix scaling (as gradient projection)
- Used in many applications (control, signal processing, machine learning)

REGULARIZED ADMM FOR QUADRATIC PROGRAMMING

Banjac, Stellato, Moehle, Goulart, Bemporad, Boyd, 2017)

• Robust "regularized" ADMM iterations:

$$\begin{array}{lcl} z^{k+1} & = & -(Q + \rho A^T A + \epsilon I)^{-1} (c - \epsilon z_k + \rho A^T (u^k - z^k)) \\ s^{k+1} & = & \min \{ \max \{ A z^{k+1} + y^k, \ell \}, u \} \\ u^{k+1} & = & u^k + A z^{k+1} - s^{k+1} \end{array}$$

- $\bullet \;\; \mbox{Works} \; \mbox{for any} \; Q \succeq 0, A, \mbox{and choice} \; \mbox{of} \; \epsilon > 0$
- Simple to code, fast, and robust
- $\bullet \ \ \text{Only needs to factorize} \left[\begin{array}{cc} Q + \epsilon I & A' \\ A & -\frac{1}{\rho}I \end{array} \right] \text{once}$
- Implemented in free osQP solver
 (Python interface: ≈ 800,000 downloads)

http://osqp.org

CAN WE SOLVE QP'S USING LEAST SQUARES?

The **least squares** (LS) problem is probably the most studied problem in numerical linear algebra



Adrien-Marie Legendre (1752–1833)



Carl Friedrich Gauss (1777–1855)

 $z^* = \arg\min \|Az - b\|_2^2$

In MATLAB: >> z=A\b

(one character!)

Nonnegative Least Squares (NNLS)

(Lawson, Hanson, 1974

$$\min_{x} ||Az - b||_{2}^{2}$$
s.t. $z > 0$

(750 chars in Embedded MATLAB)

Bounded-Variable Least Squares (BVLS)

Stark, Parker, 1995

$$\min_{z} \quad ||Az - b||_{2}^{2}$$
s.t. $\ell \le z \le u$

See Nilay's next talk

SOLVING OP'S VIA NONNEGATIVE LEAST SOUARES

(Bemporad, 2016)

Complete the squares and transform QP to least distance problem (LDP)

$$\min_{z} \quad \frac{1}{2}z'Qz + c'z$$
s.t. $Gz \le g$

$$Q = Q' > 0$$

$$Q = L'L$$

$$u \triangleq Lz + L^{-T}c$$

$$\min_{u} \quad \frac{1}{2} ||u||^2
\text{s.t.} \quad Mu \le d$$

An LDP can be solved by the NNLS (Lawson, Hanson, 1974)

$$\min_{y} \quad \frac{1}{2} \left\| \begin{bmatrix} M' \\ d' \end{bmatrix} y + \begin{bmatrix} 0 \\ 1 \end{bmatrix} \right\|_{2}^{2} \qquad M = GL^{-1}$$
 s.t. $y \ge 0$

$$M = GL^{-1}$$
$$d = b + GQ^{-1}c$$

• If residual = 0 then QP is infeasible. Otherwise set

$$z^* = -\frac{1}{1 + d'y^*} L^{-1} M'y^* - Q^{-1}c$$

Bemporad, 2018)

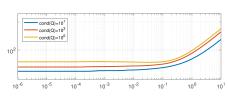
- QP solver based on NNLS is not very robust numerically
- Key idea: Solve a sequence of QP via NNLS within proximal-point iterations

$$z_{k+1} = \arg\min_{z} \quad \frac{1}{2}z'Qz + c'z + \frac{\epsilon}{2}||z - z_{k}||_{2}^{2}$$

s.t. $Az \le b$
 $Gx = g$

- Numerical robustness: $Q + \epsilon I$ can be arbitrarily well conditioned!
- Choice of ϵ is not critical

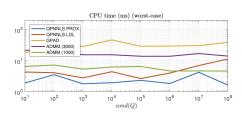
total number of active-set iterations as a function of $\boldsymbol{\epsilon}$



- Each QP is heavily warm-started and makes very few active-set changes
- Recursive LDL^T decompositions/rank-1 updates exploited for max efficiency

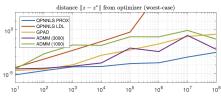
SOLVING QP'S VIA NNLS AND PROXIMAL POINT ITERATIONS

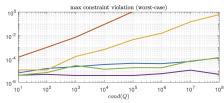
(Bemporad, 2018)



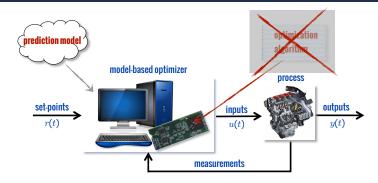
single precision arithmetic

30 vars, 100 constraints (Macbook Pro 3 GHz Intel Core i7)





MPC WITHOUT ON-LINE QP





 Can we implement constrained linear MPC without an on-line QP solver?



EXPLICIT MODEL PREDICTIVE CONTROL

Continuous & piecewise affine solution of strictly convex multiparametric QP

$$z^*(x) = \arg\min_{z} \quad \frac{1}{2}z'Qz + x'F'z$$
s.t. $Gz \le W + Sx$

(Bemporad, Morari, Dua, Pistikopoulos, 2002)



Corollary: linear MPC is continuous & piecewise affine!

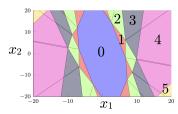
$$z^* = \begin{bmatrix} \mathbf{u_0} \\ u_1 \\ \vdots \\ u_{N-1}^* \end{bmatrix}$$

• New mpQP solver based on NNLS available (Bemporad, 2015) and included in MPC Toolbox since R2014b (Bemporad, Morari, Ricker, 1998-today)

Is explicit MPC better than on-line QP (=implicit MPC)?

COMPLEXITY CERTIFICATION FOR ACTIVE-SET QP SOLVERS

• Result: The number of iterations to solve the QP via a dual active-set method is a piecewise constant function of the parameter x



(Cimini, Bemporad, 2017)

We can **exactly** quantify how many iterations (flops) the QP solver takes in the worst-case!

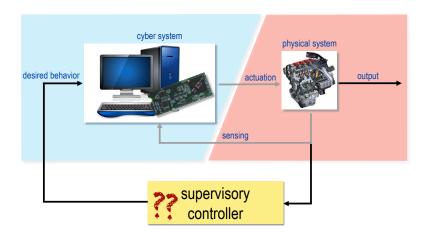
• Examples (from MPC Toolbox):

	inverted pendulum	DC motor	nonlinear demo	AFTI F16
Explicit MPC				
max flops	3382	1689	9184	16434
max memory (kB)	55	30	297	430
Implicit MPC				
max flops	3809	2082	7747	7807
sqrt	27	9	37	33
max memory (kB)	15	13	20	16

• QP certification algorithm currently used in industrial production projects

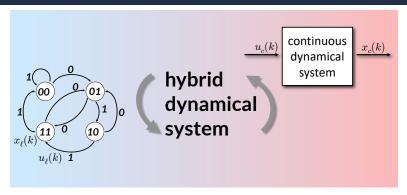


CONTROL OF CYBER-PHYSICAL SYSTEMS



Can we use MPC to synthesize a supervisory controller of a CPS?

HYBRID DYNAMICAL SYSTEMS



- Variables are binary-valued $x_{\ell} \in \{0,1\}^{n_{\ell}}, u_{\ell} \in \{0,1\}^{m_{\ell}}$
- Dynamics = finite state machine
- Logic constraints

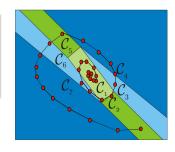
- Variables are real-valued $x_c \in \mathbb{R}^{n_c}, u_c \in \mathbb{R}^{m_c}$
- Difference/differential equations
- Linear inequality constraints

PIECEWISE AFFINE SYSTEMS

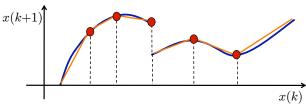
$$x(k+1) = A_{i(k)}x(k) + B_{i(k)}u(k) + f_{i(k)}$$

$$y(k) = C_{i(k)}x(k) + D_{i(k)}u(k) + g_{i(k)}$$

$$i(k) \text{ s.t. } H_{i(k)}x(k) + J_{i(k)}u(k) \le K_{i(k)}$$

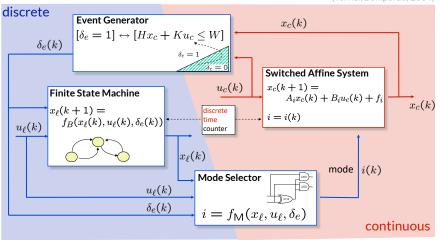


 PWA systems can approximate nonlinear dynamics arbitrarily well (even discontinuous ones)



DISCRETE HYBRID AUTOMATON (DHA)

Torrisi, Bemporad, 2004)



$$x_{\ell} \in \{0,1\}^{n_{\ell}} = ext{binary state} \qquad x_{\epsilon}$$
 $u_{\ell} \in \{0,1\}^{m_{\ell}} = ext{binary input} \qquad u_{\epsilon}$ $\delta_e \in \{0,1\}^{n_e} = ext{event variable} \qquad i \in \{0,1\}^{n_e}$

 $x_c \in \mathbb{R}^{n_c}$ = real-valued state $u_c \in \mathbb{R}^{m_c}$ = real-valued input $i \in \{1, \dots, s\}$ = current mode

TRANSFORMATION OF A DHA INTO LINEAR (IN)EQUALITIES

$$X_1 \vee X_2 = \mathsf{TRUE} \qquad \qquad \delta_1 + \delta_2 \geq 1, \qquad \delta_1, \delta_2 \in \{0,1\}$$
 Any logic statement
$$f(X) = \mathsf{TRUE} \qquad \qquad \left\{ \begin{array}{l} 1 \leq \sum\limits_{i \in P_1} \delta_i + \sum\limits_{i \in N_1} (1 - \delta_i) \\ \vdots \\ 1 \leq \sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{array} \right.$$

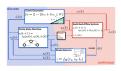
$$\left\{ \begin{array}{l} 1 \leq \sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \vdots \\ 1 \leq \sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{array} \right.$$

$$\left\{ \begin{array}{l} \sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \vdots \\ 1 \leq \sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{array} \right.$$

$$\left\{ \begin{array}{l} \sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \vdots \\ 1 \leq \sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right.$$

$$\left\{ \begin{array}{l} \left(\sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \vdots \\ \left(\sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \begin{array}{l} \left(\sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \begin{array}{l} \left(\sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \begin{array}{l} \left(\sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \begin{array}{l} \left(\sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \begin{array}{l} \left(\sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \begin{array}{l} \left(\sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \begin{array}{l} \left(\sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \begin{array}{l} \left(\sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \begin{array}{l} \left(\sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \begin{array}{l} \left(\sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \begin{array}{l} \left(\sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \begin{array}{l} \left(\sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \begin{array}{l} \left(\sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \begin{array}{l} \left(\sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \left(\sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \begin{array}{l} \left(\sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \left(\sum\limits_{i \in P_m} \delta_i + \sum \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \left(\sum\limits_{i \in P_m} \delta_i + \sum \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \left(\sum\limits_{i \in P_m} \delta_i + \sum \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \left(\sum\limits_{i \in P_m} \delta_i + \sum \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \left(\sum\limits_{i \in P_m} \delta_i + \sum \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \left(\sum\limits_{i \in P_m} \delta_i + \sum \sum\limits_{i \in N_m} (1 - \delta_i) \\ \end{bmatrix} \right. \\ \left\{ \left(\sum\limits_{i \in P_m} \delta_i + \sum \sum\limits_{i \in N_m}$$

By converting logic relations into mixed-integer linear inequalities
 a DHA can be rewritten as the Mixed Logical Dynamical (MLD) system



$$\begin{cases} x(k+1) &= Ax(k) + B_1u(k) + B_2\delta(k) + B_3z(k) + B_5 \\ y(k) &= Cx(k) + D_1u(k) + D_2\delta(k) + D_3z(k) + D_5 \\ E_2\delta(k) &+ E_3z(k) \le E_4x(k) + E_1u(k) + E_5 \end{cases}$$



$$x \in \mathbb{R}^{n_c} \times \{0, 1\}^{n_b}, \ u \in \mathbb{R}^{m_c} \times \{0, 1\}^{m_b}$$

$$y \in \mathbb{R}^{p_c} \times \{0, 1\}^{p_b}, \ \delta \in \{0, 1\}^{r_b}, \ z \in \mathbb{R}^{r_c}$$

- The translation from DHA to MLD can be automatized, see e.g. the language HYSDEL (HYbrid Systems DEscription Language) (Torrisi, Bemporad, 2004)
- MLD models allow solving MPC, verification, state estimation, and fault detection problems via mixed-integer programming

Finite-horizon optimal control problem (regulation)

$$\min \sum_{k=0}^{N-1} y_k' Q y_k + u_k' R u_k$$
s.t.
$$\begin{cases} x_{k+1} &= A x_k + B_1 u_k + B_2 \delta_k + B_3 z_k + B_5 \\ y_k &= C x_k + D_1 u_k + D_2 \delta_k + D_3 z_k + D_5 \\ E_2 \delta_k &+ E_3 z_k \le E_4 x_k + E_1 u_k + E_5 \\ x_0 &= x(t) \end{cases}$$

$$Q=Q'\succ 0, R=R'\succ 0$$

- Treat u_k, δ_k, z_k as free decision variables, $k=0,\ldots,N-1$
- Predictions can be constructed exactly as in the linear case

$$x_k = A^k x_0 + \sum_{j=0}^{k-1} A^j (B_1 u_{k-1-j} + B_2 \delta_{k-1-j} + B_3 z_{k-1-j} + B_5)$$

• After substituting x_k , y_k we get the Mixed-Integer Quadratic Programming (MIQP) problem

$$\min_{\xi} \quad \frac{1}{2}\xi'H\xi + x'(t)F'\xi + \frac{1}{2}x'(t)Yx(t)$$

s.t. $G\xi \leq W + Sx(t)$

• The optimization vector $\xi=[u_0,\ldots,u_{N-1},\delta_0,\ldots,\delta_{N-1},z_0,\ldots,z_{N-1}]$ has mixed real and binary components

 Hybrid modeling and MPC design available in Hybrid Toolbox for MATLAB (Bemporad, 2003-today)

http://cse.lab.imtlucca.it/~bemporad/hybrid/toolbox

pprox8000 downloads

pprox1.5 downloads/day

• Theorem. Let (x_r,u_r,δ_r,z_r) be the equilibrium corresponding to r. Assume x(0) such that the MIQP problem is feasible at time t=0. Then $\forall Q,R\succ 0$, $\sigma>0$ the hybrid MPC closed-loop converges asymptotically

$$\lim_{t \to \infty} y(t) = r \qquad \qquad \lim_{t \to \infty} x(t) = x_r$$

$$\lim_{t \to \infty} \delta(t) = \delta_r$$

$$\lim_{t \to \infty} z(t) = z_r$$

and all constraints are fulfilled at each time $t \geq 0$.

- The proof easily follows from standard Lyapunov arguments (see next slide)
- Lyapunov asymptotic stability and exponential stability follows if proper terminal cost and constraints are imposed (Lazar, Heemels, Weiland, Bemporad, 2006)

MIXED-INTEGER PROGRAMMING SOLVERS

- Binary constraints make Mixed-Integer Programming (MIP) a hard problem (\mathcal{NP} -complete)
- However, excellent general purpose branch & bound / branch & cut solvers available for MILP and MIQP (CPLEX, GLPK, Xpress-MP, CBC, Gurobi, ...)
- MIQP approaches tailored to embedded hybrid MPC applications:
 - B&B + (dual) active set methods for QP
 (Leyffer, Fletcher, 1998) (Axehill, Hansson, 2006) (Bemporad, 2015) (Bemporad, Naik, 2018
 - B&B + interior point methods: (Frick, Domahidi, Morari, 2015)
 - B&B + fast gradient projection: (Naik, Bemporad, 2017)
 - B&B + ADMM: (Stellato, Naik, Bemporad, Goulart, Boyd, 2018)
- No need to reach global optimum (see convergence proof)

BRANCH & BOUND METHOD FOR MIQP

• We want to solve the following MIQP

min
$$V(z) \triangleq \frac{1}{2}z'Qz + c'z$$
 $z \in \mathbb{R}^n$
s.t. $Az \leq b$ $Q = Q' \succeq 0$
 $z_i \in \{0, 1\}, \forall i \in I$ $I \subseteq \{1, \dots, n\}$

- Branch & Bound (B&B) is the simplest (and most popular) approach to solve the problem to optimality
- Key idea of B&B:
 - each binary variable $z_i, i \in I$, is either set to 0, or 1, or relaxed in [0,1]
 - solve the corresponding QP relaxation of the MIQP problem
 - use QP result to decide the next combination of fixed/relaxed variables, or to conclude that the optimal solution has been found, or that no solution exist

SOLVING MIQP VIA NNLS AND PROXIMAL-POINT ITERATIONS

Bemporad, Naik, 2018)

 Robustified approach: use NNLS + proximal-point iterations to solve QP relaxations (Bemporad, 2018)

$$z_{k+1} = \arg\min_{z} \quad \frac{1}{2}z'Qz + c'z + \frac{\epsilon}{2}||z - z_{k}||_{2}^{2}$$

s.t. $\ell \le Az \le u$
 $Gz = g$

• CPU time (ms) on MIQP coming from hybrid MPC (bm99 demo):

For $N=10$:	N	prox	-NNLS	prox-	NNLS*	GU	ROBI	CP	LEX
30 real vars 10 binary vars		avg	max	avg	max	avg	max	avg	max
,	2	2.0	2.6	2.0	2.6	1.6	2.0	3.1	6.0
$160\mathrm{inequalities}$	4	5.3	8.8	3.1	6.9	3.1	3.9	8.9	15.7
	8	29.7	71.0	8.1	43.4	7.2	13.2	15.5	80.2
$prox-NNLS^* = warm$	10	76.2	146.1	14.4	103.2	11.1	17.6	35.1	95.3
start of binary vars	12	155.8	410.8	26.9	263.4	14.9	31.2	61.7	103.7
exploited	15	484.2	1242.3	61.7	766.9	25.9	109.8	89.9	181.1

CPU time measured on Intel Core i7-4700MQ CPU 2.40 GHz

 $\bullet \;$ Consider again the MIQP problem with Hessian $Q=Q'\succ 0$

$$\min_{z} \quad V(z) \triangleq \frac{1}{2}z'Qz + c'z$$
s.t. $\ell \leq Az \leq u$

$$Gz = g$$

$$\bar{A}_{i}z \in \{\bar{\ell}_{i}, \bar{u}_{i}\}, i = 1, \dots, p$$

$$w^{k} = y^{k} + \beta_{k}(y^{k} - y^{k-1})$$

$$z^{k} = -Kw^{k} - Jx$$

$$s^{k} = \frac{1}{L}Gz^{k} - \frac{1}{L}(W + Sx)$$

$$y^{k+1} = \max\{w^{k} + s^{k}, 0\}$$

Use B&B and fast gradient projection to solve dual of QP relaxation

constraint is relaxed
$$\bar{A}_i z \leq \bar{u}_i \rightarrow y_i^{k+1} = \max \left\{ y_i^k + s_i^k, 0 \right\} \quad (y_i \geq 0)$$
 constraint is fixed $\bar{A}_i z = \bar{u}_i \rightarrow y_i^{k+1} = y_i^k + s_i^k \quad (y_i \leq 0)$ constraint is ignored $\bar{A}_i z = \bar{\ell}_i \rightarrow y_i^{k+1} = 0 \quad (y_i = 0)$

FAST GRADIENT PROJECTION FOR MIQP

Naik, Bemporad, 2017)

- Same dual QP matrices at each node, preconditioning computed only once
- Warm-start exploited, dual cost used to stop QP relaxations earlier
- Criterion based on Farkas lemma to detect QP infeasibility
- Numerical results (time in ms):

n	m	p	q	miqpGPAD	GUROBI
10	100	2	2	15.6	6.56
50	25	5	3	3.44	8.74
50	150	10	5	63.22	46.25
100	50	2	5	6.22	26.24
100	200	15	5	164.06	188.42
150	100	5	5	31.26	88.13
150	200	20	5	258.80	274.06
200	50	15	6	35.08	144.38

CPU time measured on Intel Core i7-4700MQ CPU 2.40 GHz

HEURISTIC ADMM METHOD FOR (SUBOPTIMAL) MIQP

(Takapoui, Moehle, Boyd, Bemporad, 2017)

• Consider again MIQP problem

$$\min \quad \frac{1}{2}x'Qx + q'x$$
s.t.
$$\ell \le Ax \le u$$

$$A_ix \in \{\ell_i, u_i\}, i \in I$$

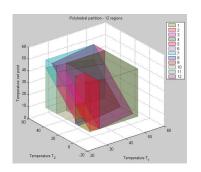
ADMM iterations:

$$\begin{array}{rcl} \text{quantization step} & \overset{+1}{=} & -(Q+\rho A^TA)^{-1}(\rho A^T(y^k-z^k)+q) \\ & z^{k+1} & = & \min\{\max\{Ax^{k+1}+y^k,\ell\},u\} \\ & z^{k+1}_i & = & \begin{cases} \ell_i & \text{if} & z^{k+1}_i < \frac{\ell_i+u_i}{2} \\ u_i & \text{if} & z^{k+1}_i \geq \frac{\ell_i+u_i}{2}, \ i \in I \end{cases} \\ & y^{k+1} & = & y^k + Ax^{k+1} - z^{k+1} \end{array}$$

- Iterations converge to a (local) solution
- Similar idea also applicable to fast gradient methods (Naik, Bemporad, 2017)

EXPLICIT HYBRID MPC

- It is possible to write hybrid MPC laws in explicit form too!
- The explicit MPC law is still piecewise affine on polyhedra



(Bemporad, Borrelli, Morari, 2000)

(Mayne, ECC 2001)

(Bemporad, Hybrid Toolbox, 2003)

(Borrelli, Baotic, Bemporad, Morari, 2005)

Alessio, Bemporad, 2006

- The control law may be discontinuous, polyhedra may overlap
- Comparison of quadratic costs can be avoided by lifting the parameter space (Fuchs, Axehill, Morari, 2015)

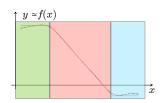
LEARNING PWA MODELS FROM DATA

• **Problem**: Given input/output pairs $\{x(k),y(k)\}$, $k=1,\ldots,N$ and number s of models, learn a piecewise affine (PWA) model $y\approx f(x)$

$$f(x) = \begin{cases} F_1 x + g_1 & \text{if } H_1 x \le K_1 \\ \vdots & \\ F_s x + g_s & \text{if } H_s x \le K_s \end{cases}$$

• Need to learn **both** the parameters $\{F_i,\,g_i\}$ of the affine submodels **and** the partition $\{H_i,\,K_i\}$ of the PWA map from data (off-line learning)

 Possibly update model and partition as new data become available (on-line learning)



APPROACHES TO PWA SYSTEM IDENTIFICATION

- Mixed-integer linear or quadratic programming (Roll, Bemporad, Ljung, 2004)
- Partition of infeasible set of inequalities (Bemporad, Garulli, Paoletti, Vicino, 2005)
- K-means clustering in a feature space (Ferrari-Trecate, Muselli, Liberati, Morari, 2003)
- Bayesian approach (Juloski, Wieland, Heemels, 2004)
- Kernel-based approaches (Pillonetto, 2016)
- Hyperplane clustering in data space (Münz, Krebs, 2002)
- Recursive multiple least squares & PWL separation (Breschi, Piga, Bemporad, 2016)

1. Estimate models $\{F_i, g_i\}$ recursively. Let $e_i(k) = y(k) - F_i x(k) - g_i$ and only update model i(k) such that

$$i(k) \leftarrow \arg\min_{i=1,\dots,s} \underbrace{e_i(k)' \Lambda_e^{-1} e_i(k)}_{\text{one-step prediction error}} + \underbrace{(x(k) - c_i)' R_i^{-1} (x(k) - c_i)}_{\text{proximity to centroid}}$$
of cluster #i

using recursive LS and inverse QR decomposition (Alexander, Ghirnikar, 1993)

This also splits the data points x(k) in clusters $C_i = \{x(k) : i(k) = i\}$

2. Compute a polyhedral partition $\{H_i,\ K_i\}$ of the regressor space via multi-category linear separation

$$\phi(x) = \max_{i=1,\dots,s} \{w_i' x - \gamma_i\}$$



PWA REGRESSION EXAMPLES

Breschi, Piga, Bemporad, 2016)

• Identification of piecewise-affine ARX model

$$\begin{split} \begin{bmatrix} y_1(k) \\ y_2(k) \end{bmatrix} &= \begin{bmatrix} -0.83 & 0.20 \\ 0.30 & -0.52 \end{bmatrix} \begin{bmatrix} y_1(k-1) \\ y_2(k-1) \end{bmatrix} + \begin{bmatrix} -0.34 & 0.45 \\ -0.30 & 0.24 \end{bmatrix} \begin{bmatrix} u_1(k-1) \\ u_2(k-1) \end{bmatrix} \\ &+ \begin{bmatrix} 0.20 \\ 0.15 \end{bmatrix} + \max \left\{ \begin{bmatrix} 0.20 & -0.90 \\ 0.10 & -0.42 \end{bmatrix} \begin{bmatrix} y_1(k-1) \\ y_2(k-1) \end{bmatrix} \right. \\ &+ \begin{bmatrix} 0.42 & 0.20 \\ 0.50 & 0.64 \end{bmatrix} \begin{bmatrix} u_1(k-1) \\ u_2(k-1) \end{bmatrix} + \begin{bmatrix} 0.40 \\ 0.30 \end{bmatrix}, \begin{bmatrix} 0 \\ 0 \end{bmatrix} \right\} + e_0(k), \end{split}$$

• Quality of fit: best fit rate (BFR) = $\max\Big\{1-\frac{\|y_{\mathrm{o},i}-\hat{y}_i\|_2}{\|y_{\mathrm{o},i}-\bar{y}_{\mathrm{o},i}\|_2},0\Big\}, i=1,2$

			N = 4000	N = 20000	N = 100000
ſ		(Off-line) RLP	96.0 %	96.5 %	99.0 %
	y_1	(Off-line) RPSN	96.2 %	96.4 %	98.9 %
		(On-line) ASGD	86.7 %	95.0 %	96.7 %
		(Off-line) RLP	96.2 %	96.9 %	99.0 %
	y_2	(Off-line) RPSN	96.3 %	96.8 %	99.0 %
		(On-line) ASGD	87.4 %	95.2 %	96.4 %

RLP = Robust linear programming

(Bennett, Mangasarian, 1994)

RPSN = Piecewise-smooth Newton method

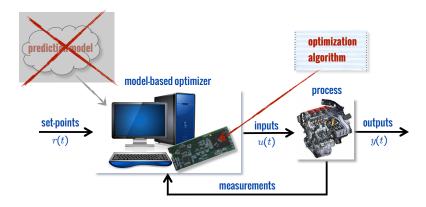
(Bemporad, Bernardini, Patrinos, 2015)

ASGD = Averaged stochastic gradient descent (Bottou, 2012)

CPU time for computing the partition:

	N = 4000	N = 20000	N = 100000
(Off-line) RLP	0.308 s	3.227 s	112.435 s
(Off-line) RPSN	0.016 s	0.086 s	0.365 s
(On-line) ASGD	0.013 s	0.023 s	0.067 s

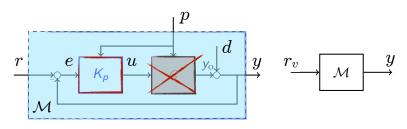
DATA-DRIVEN MPC



 Can we design an MPC controller without first identifying a model of the open-loop process?

DATA-DRIVEN DIRECT CONTROLLER SYNTHESIS

(Campi, Lecchini, Savaresi, 2002) (Formentin et al., 2015)

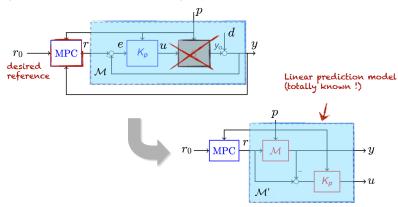


- Collect a set of data $\{u(t), y(t), p(t)\}, t = 1, \dots, N$
- $\bullet \;$ Specify a desired closed-loop linear model ${\mathcal M}$ from r to y
- Compute $r_v(t) = \mathcal{M}^\# y(t)$ from pseudo-inverse model $\mathcal{M}^\#$ of \mathcal{M}
- Identify linear (LPV) model K_p from $e_v = r_v y$ (virtual tracking error) to u

DATA-DRIVEN MPC

ullet Design a linear MPC (reference governor) to generate the reference r

(Bemporad, Mosca, 1994) (Gilbert, Kolmanovsky, Tan, 1994)



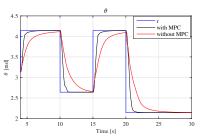
• MPC designed to handle input/output constraints and improve performance

(Piga, Formentin, Bemporad, 2017

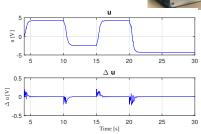
DATA-DRIVEN MPC - AN EXAMPLE

 \bullet Experimental results: MPC handles soft constraints on $u,\Delta u$ and y (motor equipment by courtesy of TU Delft)





desired tracking performance achieved



constraints on input increments satisfied

No open-loop process model is identified to design the MPC controller!

- Can we choose \mathcal{M} from data so that K_p is an optimal controller?
- **Idea**: parameterize desired closed-loop model $\mathcal{M}(\theta)$ and optimize

$$\min_{\theta} J(\theta) = \frac{1}{N} \sum_{t=0}^{N-1} \underbrace{W_y(r(t) - y_p(\theta, t))^2 + W_{\Delta u} \Delta u_p^2(\theta, t)}_{\text{performance index}} + \underbrace{W_{\text{fit}}(u(t) - u_v(\theta, t))^2}_{\text{identification error}}$$

 $\bullet \;$ Evaluating $J(\theta)$ requires synthesizing $K_p(\theta)$ from data and simulating the nominal model and control law

$$y_p(\theta, t) = \mathcal{M}(\theta)r(t) \qquad u_p(\theta, t) = K_p(\theta)(r(t) - y_p(\theta, t))$$
$$\Delta u_p(\theta, t) = u_p(\theta, t) - u_p(\theta, t - 1)$$

• Optimal θ obtained by solving a (non-convex) nonlinear programming problem

Results: linear process

$$G(z) = \frac{z - 0.4}{z^2 + 0.15z - 0.325}$$

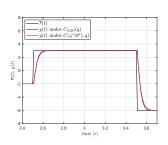
The data-driven controller is **only 1.3% worse** than model-based LQR

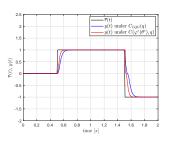
• Results: nonlinear (Wiener) process

$$y_L(t) = G(z)u(t)$$

 $y(t) = |y_L(t)| \arctan(y_L(t))$

The data-driven controller is 24% better than LQR based on identified open-loop model!





ONGOING RESEARCH ON LEARNING MPC FROM DATA

- Goal: learn MPC law from data that optimizes a given index
- Q-learning: learn Q-function defining the MPC law from data (Gros, Zanon, in press) (Zanon, Gros, Bemporad, 2019)
 - Policy gradient methods: learn optimal policy coefficients directly from data using stochastic gradient descent (Ferrarotti, Bemporad, 2019)
- Global optimization methods: learn MPC parameters (weights, models, horizon, solver tolerances, ...) by optimizing observed closed-loop performance (Piga, Forgione, Formentin, Bemporad, 2019) (Forgione, Piga, Bemporad, in preparation)
- Lessons learned so far: if chosen model/policy structure does not include real plant/optimal policy
 - optimal policy learned from data can be better than model-based optimal policy
 - when open-loop model is used as a tuning parameter, learned model can be quite different from best open-loop model that can be identified from the same data

CONCLUSIONS

- MPC is a universal control methodology to provide autonomy to many CPS's:
 - different models (linear, nonlinear, hybrid, stochastic, ...)
 - optimize closed-loop performance subject to constraints
 - widely applicable to many industrial sectors

• MPC research:

- 1. Linear, uncertain, explicit, hybrid, nonlinear MPC: mature theory
- 2. Stochastic MPC, economic MPC: still open issues
- 3. Embedded optimization methods for MPC: still room for many new ideas
- 4. Data-driven MPC: a lot of open issues. There is a lot to "learn" from machine learning
- MPC technology: is MPC mature for widespread use in industrial applications?

MPC IN AUTOMOTIVE PRODUCTION

The MPC developed by **General Motors** and **ODYS** for torque tracking in turbocharged gasoline engines is in high-volume production since 2018

Multivariable system, 4 inputs, 4 outputs.
 QP solved in real time on ECU

(Bemporad, Bernardini, Long, Verdejo, 2018)

 Supervisory MPC for powertrain control also in production since 2018 (Bemporad, Bernardini, Livshiz, Pattipati, 2018)



First known mass production of MPC in the automotive industry

http://www.odys.it/odys-and-gm-bring-online-mpc-to-production

